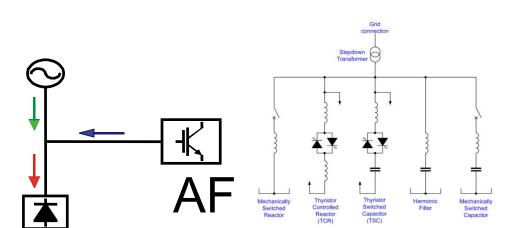
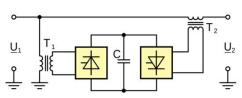
Grid Connected Power Electronics

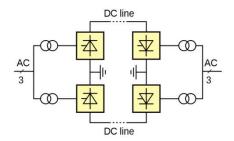
Acronyms

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- APF Active Power Filter
- UPFC Unified Power Flow Controller
- SVC Static Var Converter
- HVDC High Voltage Direct Current







All made to improve "Power Quality"

Non ideal loads

• are loads that:

- are non-resistive -> consume reactive power
- vary with time or phase -> consume harmonic current components.
- are different in different phases
 -> consume negative sequence
 currents

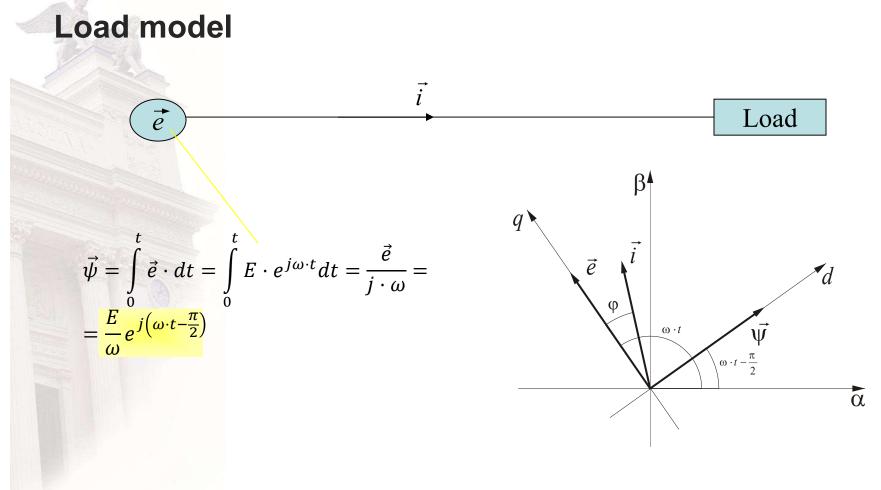
Ways to improve the loads

Self improvement

 Solutions that draw as "ideal" current as possible from the grid

Compensation

 A parallel unit is used to counteract the non ideal currents drawn by the main load



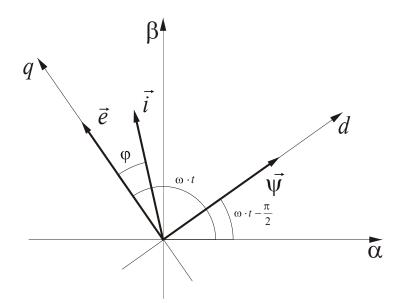
Reactive power

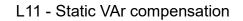
• A phase lag between the voltage and the current:

 $\int_{2}^{3} \cdot \hat{\imath} \cdot e^{j(\omega t - \phi)}$

 $\sqrt{\frac{3}{2}} \cdot \hat{\imath} \cdot e^{j\left(\frac{\pi}{2} - \phi\right)}$

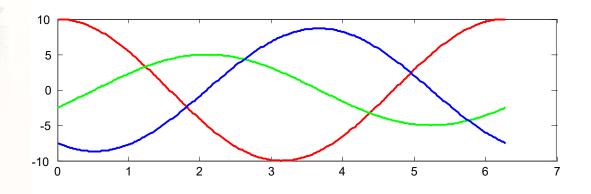
In flux coordinates:





Assymetric load

 The phase currents are not equal in amplitude or phase lag ...

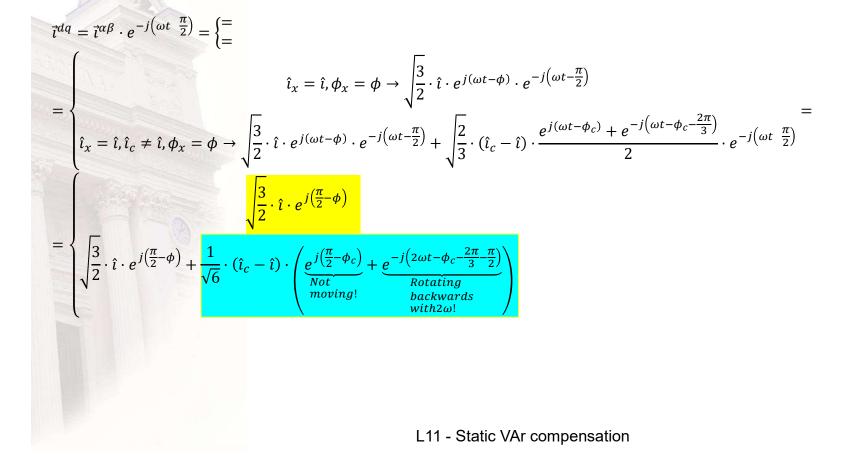


L11 - Static VAr compensation

An assymetric load current vector in the (α , β)-frame

$$\vec{l} = \sqrt{\frac{2}{3}} \cdot \left(\hat{l}_{a} \cdot \frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} \cdot 1 + \hat{l}_{b} \cdot \frac{e^{j(\omega t - \phi_{b} - \frac{2\pi}{3})} + e^{-j(\omega t - b - \frac{2\pi}{3})}}{2} \cdot e^{j\frac{2\pi}{3}} + \hat{l}_{c} \cdot \frac{e^{j(\omega t - \phi_{c} - \frac{4\pi}{3})} + e^{-j(\omega t - \phi_{c} - \frac{4\pi}{3})}}{2} \cdot e^{j\frac{4\pi}{3}} \right) = \\ = \sqrt{\frac{2}{3}} \cdot \left(\hat{l}_{a} \cdot \frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} + \hat{l}_{b} \cdot \frac{e^{j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b} - \frac{4\pi}{3})}}{2} + \hat{l}_{c} \cdot \frac{e^{j(\omega t - \phi_{c})} + e^{-j(\omega t - \phi_{c} - \frac{2\pi}{3})}}{2} \right) = \\ = \begin{cases} \hat{l}_{x} = \hat{l}_{x} \phi_{x} = \phi \rightarrow \sqrt{\frac{2}{3}} \cdot \frac{3 \cdot \hat{l}}{2} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{2}{3}} \cdot \frac{3 \cdot \hat{l}}{2} \left(\frac{e^{-j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{c} - \frac{4\pi}{3})}}{2} + e^{-j(\omega t - \phi_{c} - \frac{2\pi}{3})} \right) = \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} \\ \hat{l}_{x} = \hat{l}_{x} \hat{l}_{c} \neq \hat{l}_{y} \phi_{x} = \phi \rightarrow \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{2}{3}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \frac{2}{\sqrt{\frac{3}{3}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)}}{2} \right) = \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \frac{2\pi}{3} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \frac{2\pi}{3} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{l} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}$$

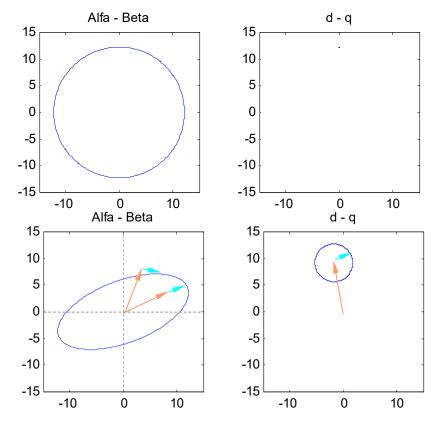
An assymetric load current vector in the (d,q)-frame



M-file "Load current vectors" - demo

Symmetric load

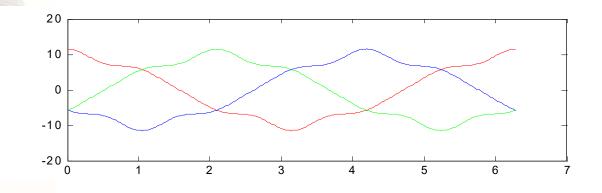
Assymetric



L11 - Static VAr compensation

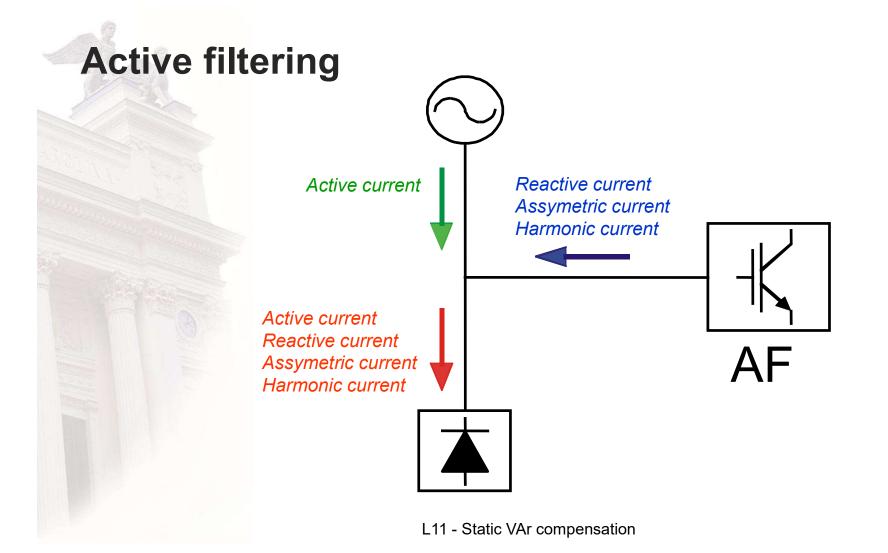
Harmonics

Non linear load impedance

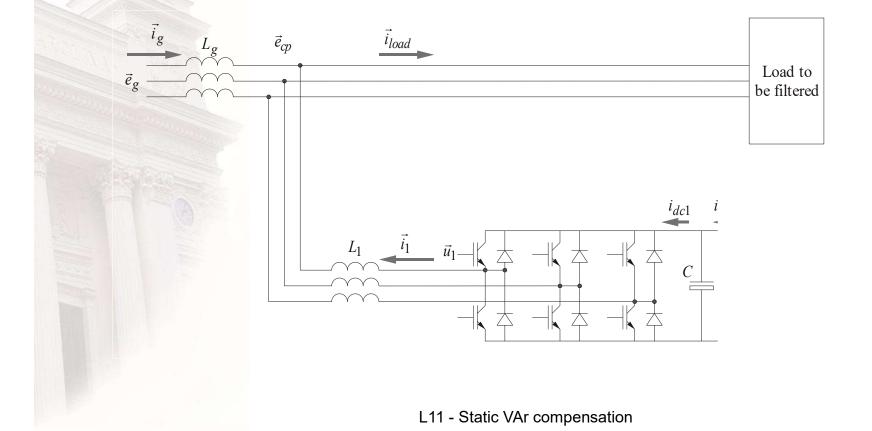


5'th and 7'th harmonic example with Simulink

$$\vec{\imath}^{\alpha\beta} = \sqrt{\frac{3}{2}} \cdot \hat{\imath}_1 \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{\imath}_5 \cdot e^{j(-5\omega t - \cdot)} + \sqrt{\frac{3}{2}} \cdot \hat{\imath}_7 \cdot e^{j(7\omega t - \phi)}$$
$$\vec{\imath}^{dq} = \sqrt{\frac{3}{2}} \cdot \hat{\imath}_1 \cdot e^{j(\frac{\pi}{2} - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{\imath}_5 \cdot e^{j(-6\omega t - \cdot)} + \sqrt{\frac{3}{2}} \cdot \hat{\imath}_7 \cdot e^{j(6\omega t - \phi)}$$
L11 - Static VAr compensation



Shunt Active Filter

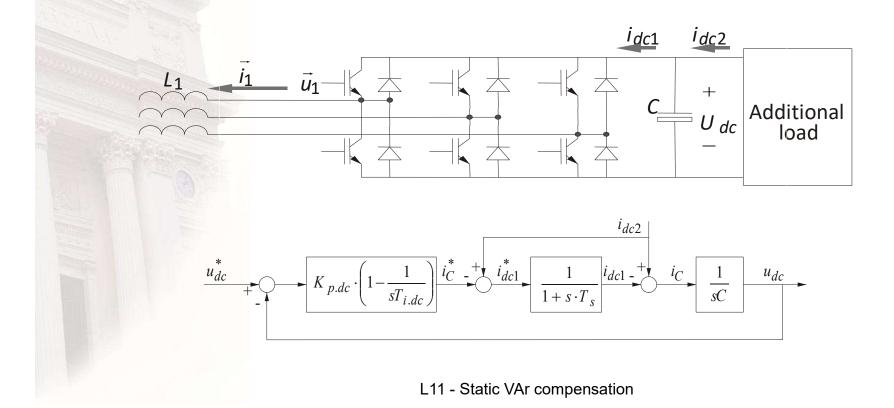


AC side Current Control

Vector Control with Field
 Orientation

$$\vec{u}_{1}^{*}(k) = \left(\frac{L_{1}}{T_{s}} + \frac{R_{1}}{2}\right) \cdot \left(\left(\vec{t}_{1}^{*}(k) - \hat{\vec{t}}_{1}(k)\right) + \frac{T_{s}}{\left(\frac{L}{R} + \frac{T_{s}}{2}\right)} \cdot \sum_{n=0}^{n=k-1} \left(\vec{t}_{1}^{*}(n) - \hat{\vec{t}}_{1}(n)\right)\right) + \hat{\vec{e}}_{cp}(k)$$

DC link Voltage Control System



Controller Parameters ...

Use Symmetric Optimum

$$\zeta = \frac{a-1}{2}$$

$$T_{i.dc} = a^2 \cdot T_s$$
, where $a > 1$

$$K_{p.dc} = \frac{a \cdot C}{T_{i.dc}}$$

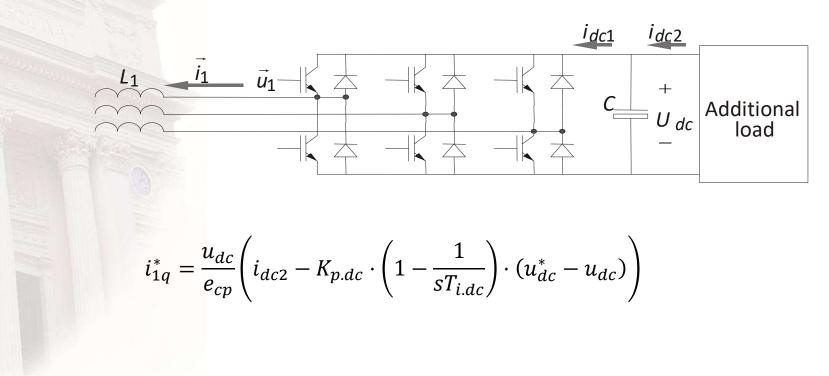
Convert DC to AC current references

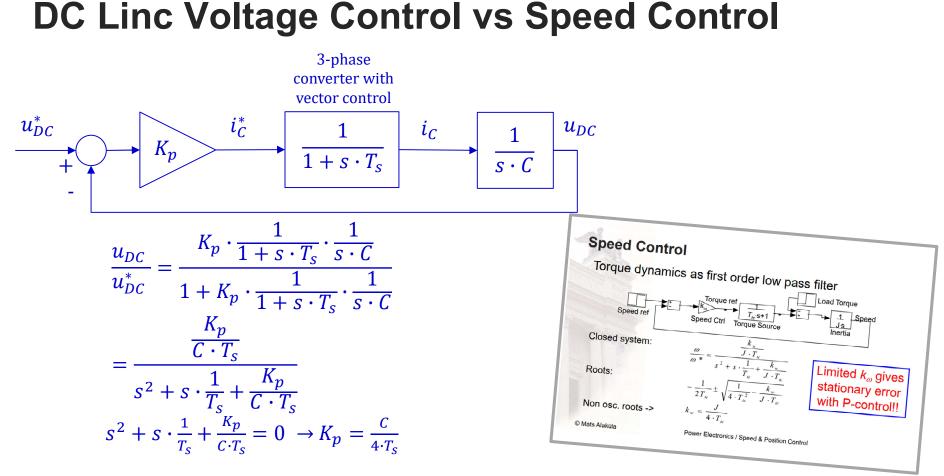
$$p(t) = Ri_{1d}^2 + Ri_{1q}^2 + L\frac{di_{1d}}{dt}i_{1d} + L\frac{di_{1q}}{dt}i_{1q} + e_{cp,q}i_{1q} = u_{dc} \cdot i_{dc1} \approx e_{cp,q}i_{1q}$$

$$\downarrow$$

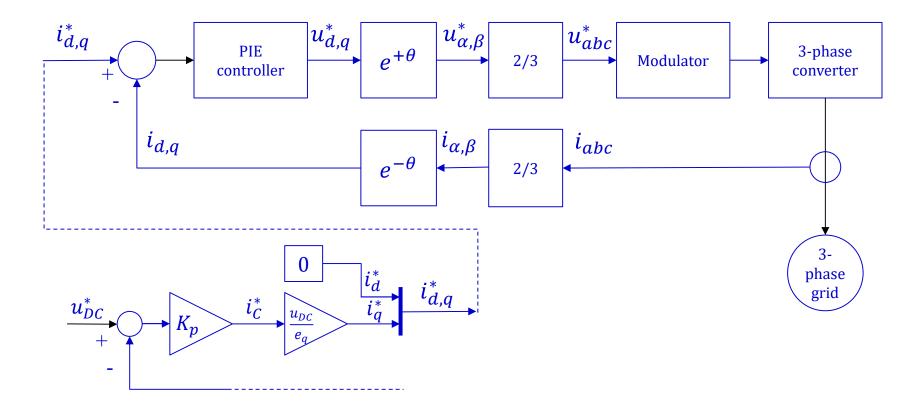
$$i_{dc1} = \frac{e_{cp,q}}{u_{dc}} \cdot i_{1q}$$

DC link voltage controller





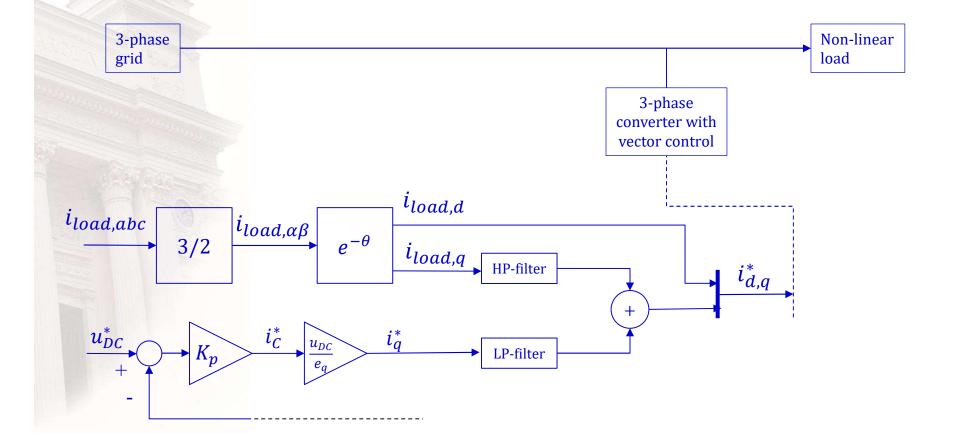




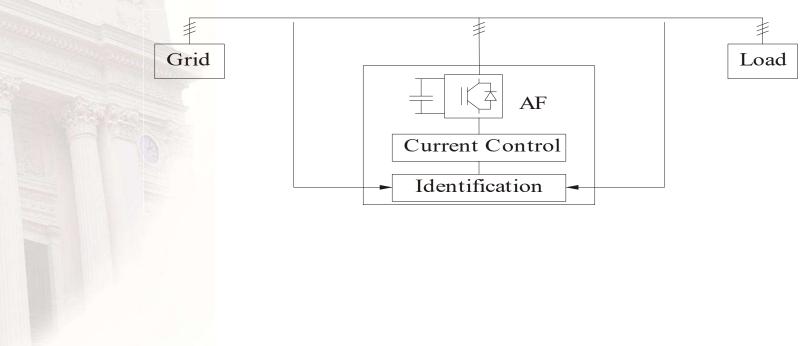
The 3phase Grid connected converter is the DC link current source

L11 - Static VAr compensation

The full control system

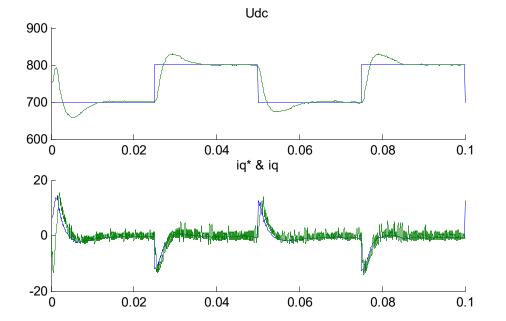


Active filter control



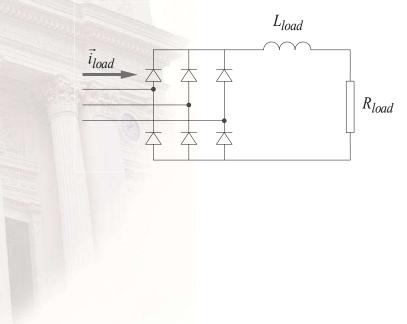
Example of DC voltage control

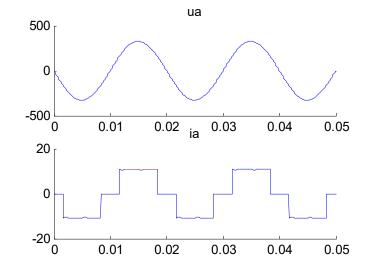
>> L=0.01; >> R=1; >> Ts=0.0005; >> Tidc=9*Ts; >> Kpdc=3*Cdc/Tidc; >> Cdc=1e-4;



L11 - Static VAr compensation

Example with active filtering





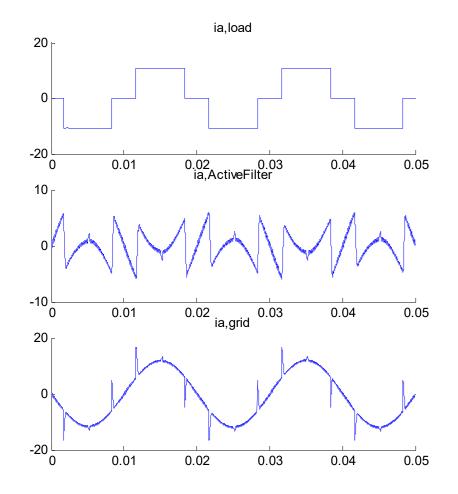
L11 - Static VAr compensation

Filter Current References

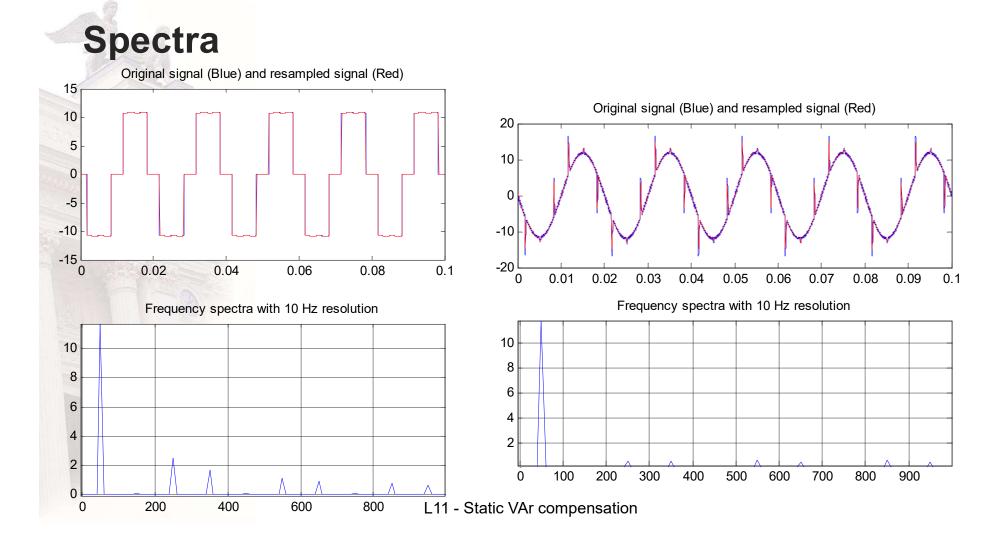
 $i_{d,ActiveFilter}^* = i_{d,load}$ $i_{q,ActiveFilter}^{*} = i_{q,load} \cdot \frac{s \cdot T_{f}}{1 + s \cdot T_{f}} + \frac{u_{dc}}{e_{cp}} \left(i_{dc2} - K_{p.dc} \cdot \left(1 - \frac{1}{sT_{i.dc}} \right) \cdot \left(u_{dc}^{*} - u_{dc} \right) \right) \cdot \frac{1}{1 + s \cdot T_{f}}$

Filter Currents

>> L=0.01; >> R=1; >> Ts=0.00005; >> Rload=50; >> Lload=0.1; >> Tf=10e-3; >> Tidc=9*Tf; >> Kpdc=3*Cdc/Tidc;



L11 - Static VAr compensation



Exercises

- 3-phase vectors
 1.11, 1.12, 1.13, 1,14
- 3-phase current control
 - 2.4